

**OBSOLETE PRODUCT  
POSSIBLE SUBSTITUTE PRODUCT  
HGTG30N60B3D**

**54A, 600V, Rugged UFS Series N-Channel IGBT with Anti-Parallel Ultrafast Diode**

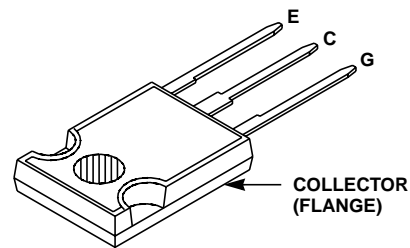
This IGBT was designed for optimum performance in the demanding world of motor control operation as well as other high voltage switching applications. This device demonstrates RUGGED performance capability when subjected to harsh SHORT CIRCUIT WITHSTAND TIME (SCWT) conditions. The parts have ULTRAFast (UFS) switching speed while the on-state conduction losses have been kept at a low level.

**Features**

- 54A, 600V,  $T_C = 25^\circ\text{C}$
- 600V Switching SOA Capability
- Typical Fall Time at  $T_J = 150^\circ\text{C}$  . . . . . 200ns
- Short Circuit Rating at  $T_J = 150^\circ\text{C}$  . . . . . 10 $\mu\text{s}$
- Low Conduction Loss

**Package**

JEDEC STYLE TO-247

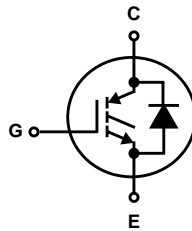


**Ordering Information**

PART NUMBER	PACKAGE	BRAND
HGTG27N60C3DR	TO-247	27N60C3DR

NOTE: When ordering, use the entire part number.

**Symbol**



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**INTERSIL CORPORATION IGBT PRODUCT IS COVERED BY ONE OR MORE OF THE FOLLOWING U.S. PATENTS**

4,364,073	4,417,385	4,430,792	4,443,931	4,466,176	4,516,143	4,532,534	4,587,713
4,598,461	4,605,948	4,620,211	4,631,564	4,639,754	4,639,762	4,641,162	4,644,637
4,682,195	4,684,413	4,694,313	4,717,679	4,743,952	4,783,690	4,794,432	4,801,986
4,803,533	4,809,045	4,809,047	4,810,665	4,823,176	4,837,606	4,860,080	4,883,767
4,888,627	4,890,143	4,901,127	4,904,609	4,933,740	4,963,951	4,969,027	

## HGTG27N60C3DR

### Absolute Maximum Ratings $T_C = 25^\circ$ Unless Otherwise Specified

	HGTG27N60C3DR	UNITS
Collector to Emitter Voltage . . . . .	600	V
Collector Current Continuous		
At $T_C = 25^\circ\text{C}$ . . . . .	54	A
At $T_C = 110^\circ\text{C}$ . . . . .	27	A
Collector Current Pulsed (Note 1) . . . . .	108	A
Gate to Emitter Voltage Continuous . . . . .	$\pm 20$	V
Gate to Emitter Voltage Pulsed . . . . .	$\pm 30$	V
Switching Safe Operating Area at $T_J = 150^\circ\text{C}$ (Figure 12) . . . . .	108A at 600V	
Power Dissipation Total at $T_C = 25^\circ\text{C}$ . . . . .	208	W
Power Dissipation Derating $T_C > 25^\circ\text{C}$ . . . . .	1.67	W/ $^\circ\text{C}$
Reverse Voltage Avalanche Energy . . . . .	100	mJ
Operating and Storage Junction Temperature Range . . . . .	-40 to 150	$^\circ\text{C}$
Maximum Lead Temperature for Soldering . . . . .	260	$^\circ\text{C}$
Short Circuit Withstand Time (Note 2) at $V_{GE} = 15\text{V}$ . . . . .	10	$\mu\text{s}$

*CAUTION: Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.*

#### NOTES:

- Pulse width limited by maximum junction temperature.
- $V_{CE(PK)} = 440\text{V}$ ,  $T_J = 150^\circ\text{C}$ ,  $R_G = 3\Omega$ .

### Electrical Specifications $T_C = 25^\circ\text{C}$ , Unless Otherwise Specified

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNITS	
Collector to Emitter Breakdown Voltage	$BV_{CES}$	$I_C = 250\mu\text{A}$ , $V_{GE} = 0\text{V}$	600	-	-	V	
Emitter to Collector Breakdown Voltage	$BV_{ECS}$	$I_C = 10\text{mA}$ , $V_{GE} = 0\text{V}$	15	-	-	V	
Collector to Emitter Leakage Current	$I_{CES}$	$V_{CE} = BV_{CES}$	$T_C = 25^\circ\text{C}$	-	-	250	$\mu\text{A}$
			$T_C = 150^\circ\text{C}$	-	-	3.0	mA
Collector to Emitter Saturation Voltage	$V_{CE(SAT)}$	$I_C = I_{C110}$ , $V_{GE} = 15\text{V}$	$T_C = 25^\circ\text{C}$	-	1.8	2.2	V
			$T_C = 150^\circ\text{C}$	-	2.1	2.5	V
Gate to Emitter Threshold Voltage	$V_{GE(TH)}$	$I_C = 250\mu\text{A}$ , $V_{CE} = V_{GE}$	3.5	5.7	7.5	V	
Gate to Emitter Leakage Current	$I_{GES}$	$V_{GE} = \pm 20\text{V}$	-	-	$\pm 100$	nA	
Switching SOA (See Figure 12)	SSOA	$T_J = 150^\circ\text{C}$ , $R_G = 3\Omega$ , $L = 50\mu\text{H}$ $V_{GE} = 15\text{V}$ , $V_{CE(PK)} = 600\text{V}$	108	-	-	A	
Gate to Emitter Plateau Voltage	$V_{GEP}$	$I_C = I_{C110}$ , $V_{CE} = 0.5 BV_{CES}$	-	9.0	-	V	
On-State Gate Charge	$Q_{G(ON)}$	$I_C = I_{C110}$ , $V_{CE} = 0.5 BV_{ES}$	$V_{GE} = 15\text{V}$	-	156	203	nC
			$V_{GE} = 20\text{V}$	-	212	277	nC
Current Turn-On Delay Time	$t_{d(ON)I}$	$T_J = 150^\circ\text{C}$ $I_{CE} = I_{C110}$ $V_{CE(PK)} = 0.8 BV_{CES}$ $V_{GE} = 15\text{V}$ $R_G = 3\Omega$ $L = 1\text{mH}$  Diode Used in Test Circuit RURP3060 at $150^\circ\text{C}$	-	38	-	ns	
Current Rise Time	$t_{rI}$		-	30	-	ns	
Current Turn-Off Delay Time	$t_{d(OFF)I}$		-	250	500	ns	
Current Fall Time	$t_{fI}$		-	200	400	ns	
Turn-Off Voltage $dv/dt$ (Note 3)	$dV_{CE}/dt$		-	2	-	V/ns	
Turn-On Voltage $dv/dt$ (Note 3)	$dV_{CE}/dt$		-	7	-	V/ns	
Turn-On Energy (Note 4)	$E_{ON}$		-	2.3	-	mJ	
Turn-Off Energy (Note 5)	$E_{OFF}$		-	2.0	-	mJ	
Diode Forward Voltage	$V_{EC}$	$I_{EC} = 27\text{A}$	-	-	1.5	V	

## HGTG27N60C3DR

**Electrical Specifications**  $T_C = 25^\circ\text{C}$ , Unless Otherwise Specified (Continued)

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNITS
Diode Reverse Recovery Time	$t_{rr}$	$I_{EC} = 1\text{A}$ , $dI_{EC}/dt = 200\text{A}/\mu\text{s}$	-	-	55	ns
		$I_{EC} = 27\text{A}$ , $dI_{EC}/dt = 200\text{A}/\mu\text{s}$	-	-	60	ns
Thermal Resistance	$R_{\theta JC}$	IGBT	-	-	0.6	$^\circ\text{C}/\text{W}$
		Diode	-	-	1.25	$^\circ\text{C}/\text{W}$

## NOTES:

- $dV_{CE}/dt$  depends on the diode used and the temperature of the diode.
- Turn-On Energy Loss ( $E_{ON}$ ) includes losses due to the diode recovery and is defined as the integral of the instantaneous power loss starting at the leading edge of the input pulse and ending at the point where the collector voltage equals  $V_{CE(SAT)}$ . This value of  $E_{ON}$  was obtained with a RURP3060 diode at  $T_J = 150^\circ\text{C}$ . A different diode or temperature will result in a different  $E_{ON}$ . For example, with diode at  $T_J = 25^\circ\text{C}$   $E_{ON}$  is about one half the value at  $150^\circ\text{C}$ .
- Turn-Off Energy Loss ( $E_{OFF}$ ) is defined as the integral of the instantaneous power loss starting at the trailing edge of the input pulse and ending at the point where the collector current equals zero ( $I_{CE} = 0\text{A}$ ). All devices were tested per JEDEC standard No. 24-1 Method for measurement of power device turn-off switching loss. This test method produces the true total Turn-Off Energy Loss.

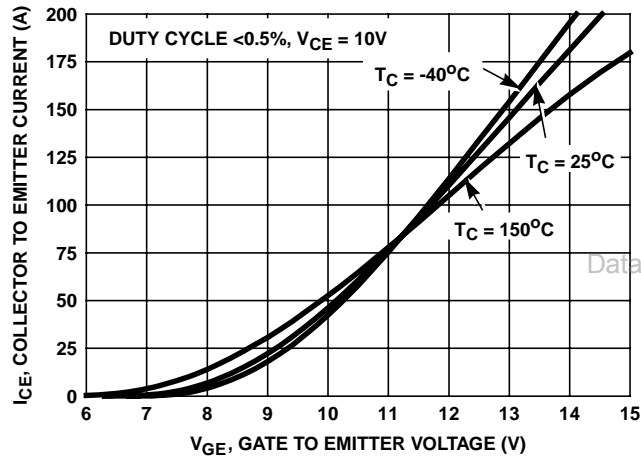
**Typical Performance Curves**

FIGURE 1. TRANSFER CHARACTERISTICS

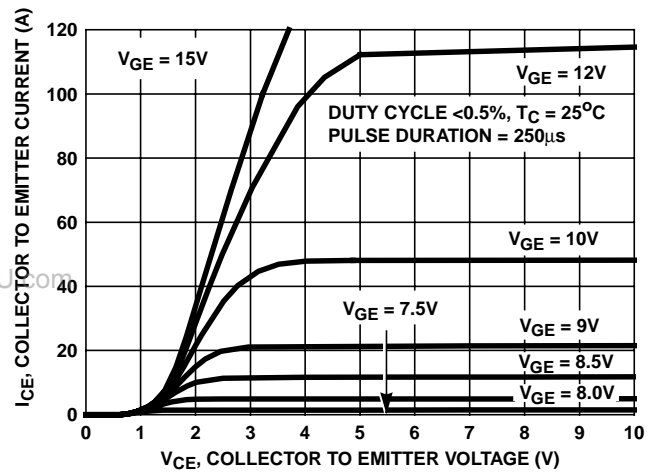


FIGURE 2. SATURATION CHARACTERISTICS

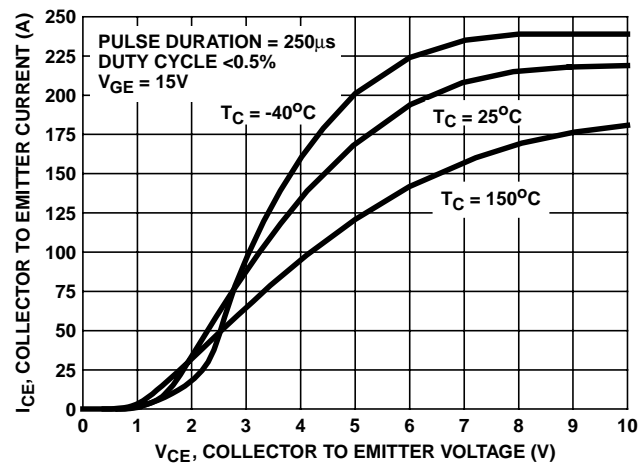


FIGURE 3. COLLECTOR TO EMITTER ON STATE VOLTAGE

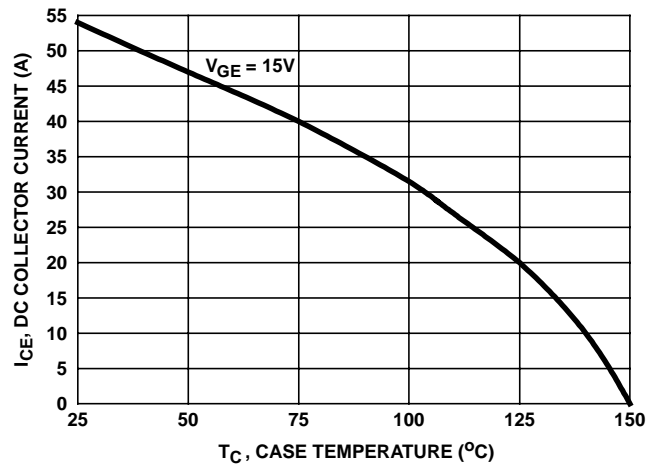


FIGURE 4. DC COLLECTOR CURRENT vs CASE TEMPERATURE

### Typical Performance Curves (Continued)

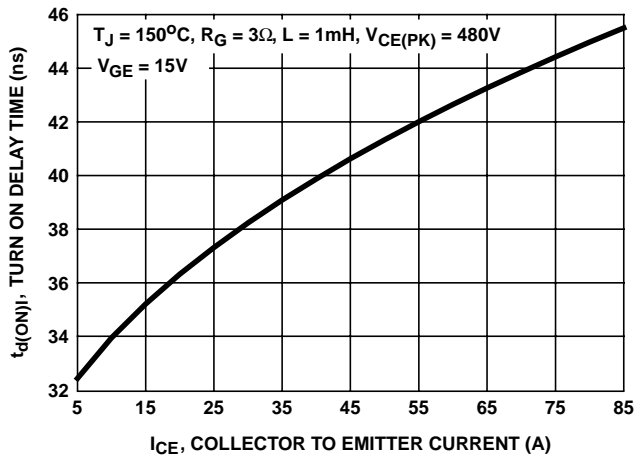


FIGURE 5. TURN ON DELAY TIME vs COLLECTOR TO EMITTER CURRENT

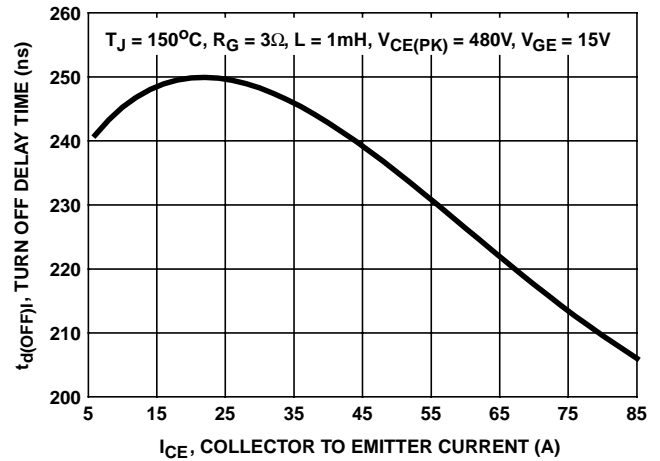


FIGURE 6. TURN OFF DELAY TIME vs COLLECTOR TO EMITTER CURRENT

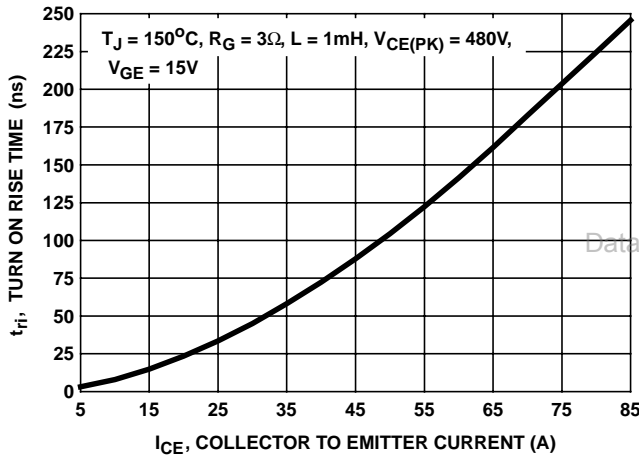


FIGURE 7. TURN ON RISE TIME vs COLLECTOR TO EMITTER CURRENT

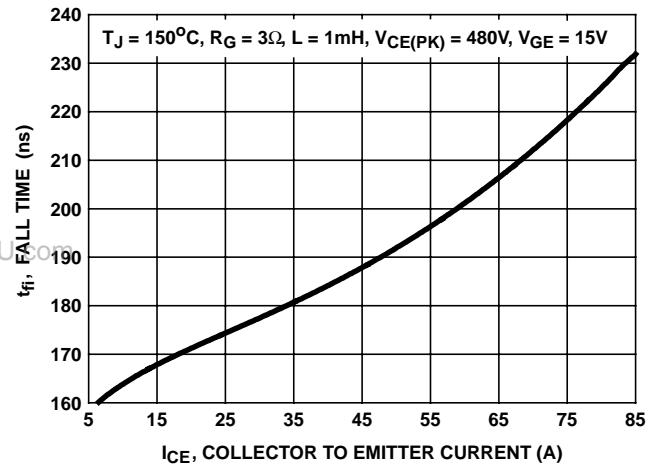


FIGURE 8. TURN OFF FALL TIME vs COLLECTOR TO EMITTER CURRENT

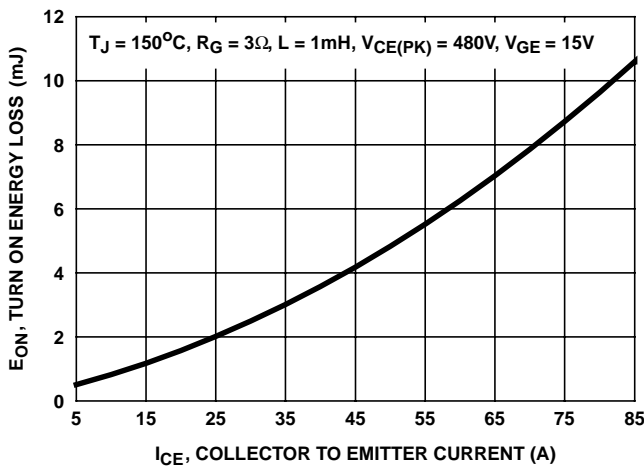


FIGURE 9. TURN ON ENERGY LOSS vs COLLECTOR TO EMITTER CURRENT

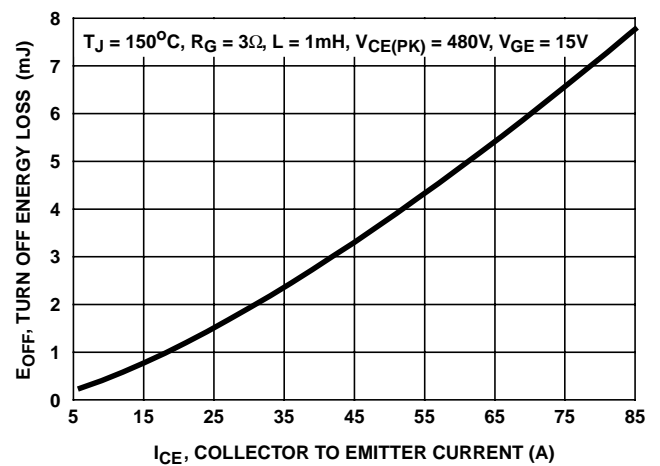


FIGURE 10. TURN OFF ENERGY LOSS vs COLLECTOR TO EMITTER CURRENT

Typical Performance Curves (Continued)

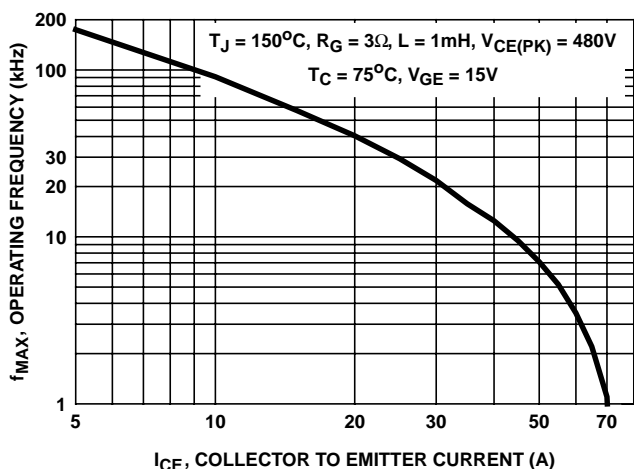


FIGURE 11. OPERATING FREQUENCY vs COLLECTOR TO EMITTER CURRENT

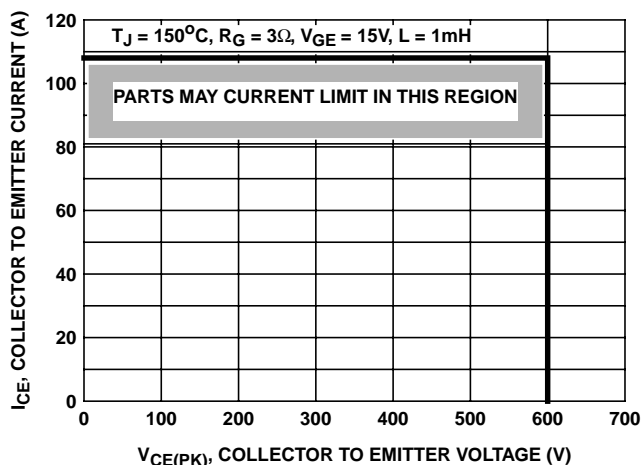


FIGURE 12. SWITCHING SAFE OPERATING AREA

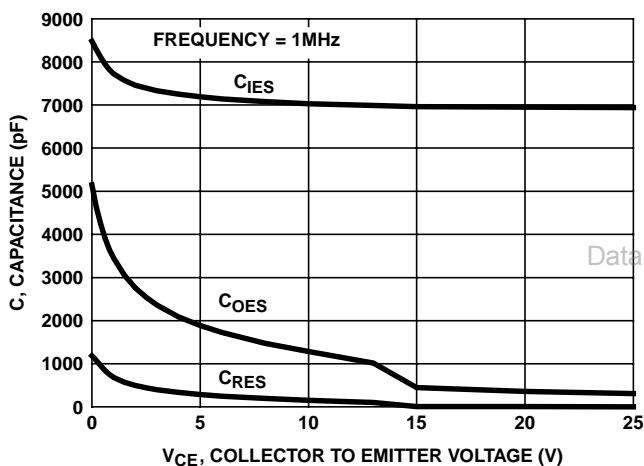


FIGURE 13. CAPACITANCE vs COLLECTOR TO EMITTER VOLTAGE

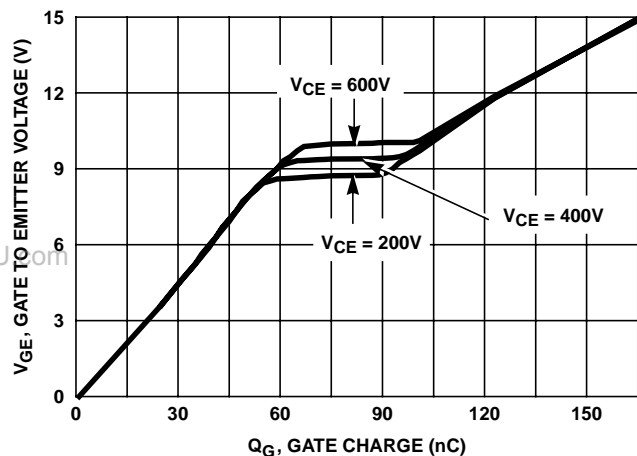


FIGURE 14. GATE CHARGE WAVEFORMS

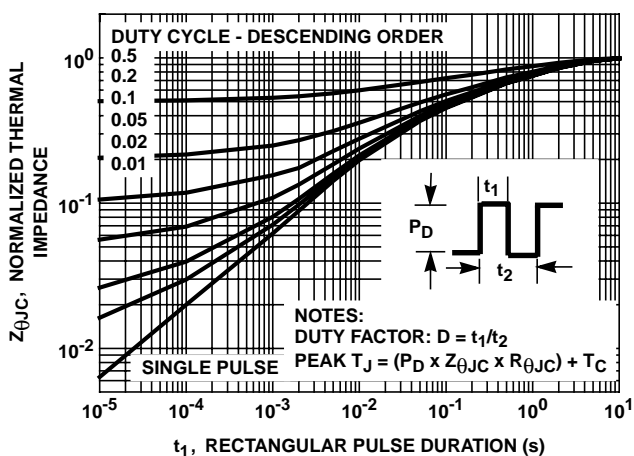


FIGURE 15. IGBT NORMALIZED TRANSIENT THERMAL IMPEDANCE, JUNCTION TO CASE

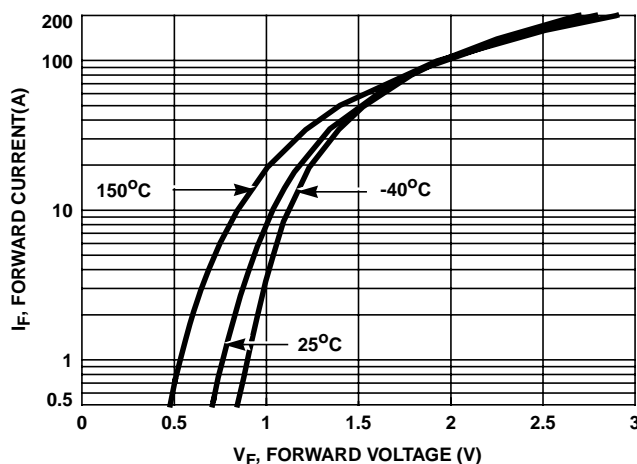


FIGURE 16. DIODE FORWARD CURRENT vs FORWARD VOLTAGE DROP

# HGTG27N60C3DR

## Typical Performance Curves (Continued)

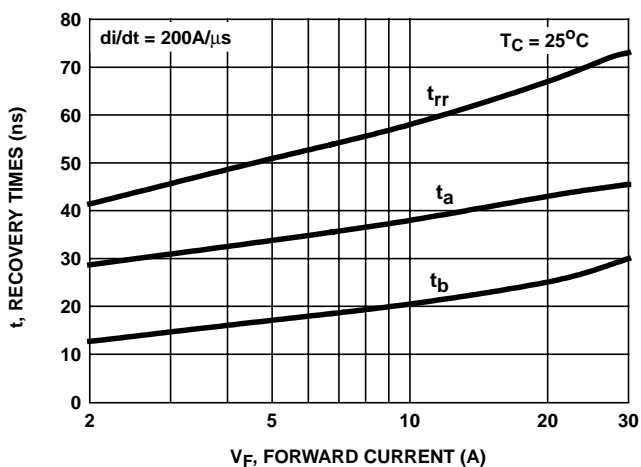


FIGURE 17. RECOVERY TIMES vs FORWARD CURRENT

## Test Circuit and Waveforms

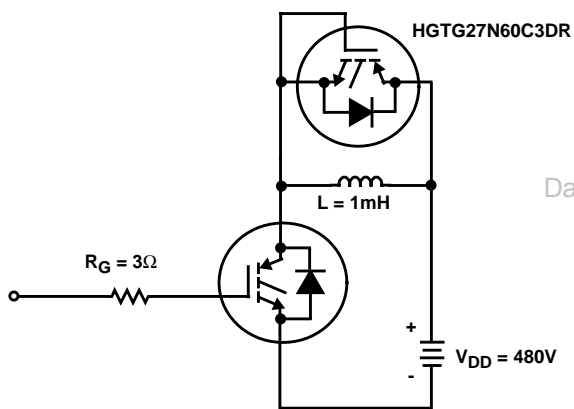


FIGURE 18. INDUCTIVE SWITCHING TEST CIRCUIT

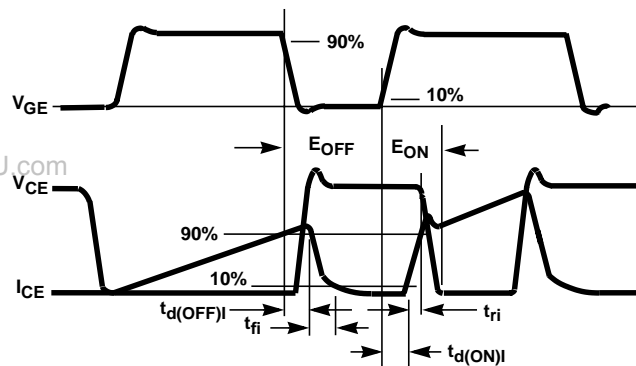


FIGURE 19. SWITCHING TEST WAVEFORMS

## HGTG27N60C3DR

### Handling Precautions for IGBTs

Insulated Gate Bipolar Transistors are susceptible to gate-insulation damage by the electrostatic discharge of energy through the devices. When handling these devices, care should be exercised to assure that the static charge built in the handler's body capacitance is not discharged through the device. With proper handling and application procedures, however, IGBTs are currently being extensively used in production by numerous equipment manufacturers in military, industrial and consumer applications, with virtually no damage problems due to electrostatic discharge. IGBTs can be handled safely if the following basic precautions are taken:

1. Prior to assembly into a circuit, all leads should be kept shorted together either by the use of metal shorting springs or by the insertion into conductive material such as "ECCOSORBD™ LD26" or equivalent.
2. When devices are removed by hand from their carriers, the hand being used should be grounded by any suitable means - for example, with a metallic wristband.
3. Tips of soldering irons should be grounded.
4. Devices should never be inserted into or removed from circuits with power on.
5. **Gate Voltage Rating** - Never exceed the gate-voltage rating of  $V_{GEM}$ . Exceeding the rated  $V_{GE}$  can result in permanent damage to the oxide layer in the gate region.
6. **Gate Termination** - The gates of these devices are essentially capacitors. Circuits that leave the gate open-circuited or floating should be avoided. These conditions can result in turn-on of the device due to voltage buildup on the input capacitor due to leakage currents or pickup.
7. **Gate Protection** - These devices do not have an internal monolithic Zener diode from gate to emitter. If gate protection is required an external Zener is recommended.

### Operating Frequency Information

Operating frequency information for a typical device (Figure 11) is presented as a guide for estimating device performance for a specific application. Other typical frequency vs collector current ( $I_{CE}$ ) plots are possible using the information shown for a typical unit in Figures 3, 5, 6, 9 and 10. The operating frequency plot (Figure 11) of a typical device shows  $f_{MAX1}$  or  $f_{MAX2}$  whichever is smaller at each point. The information is based on measurements of a typical device and is bounded by the maximum rated junction temperature.

$f_{MAX1}$  is defined by  $f_{MAX1} = 0.05 / (t_{D(OFF)I} + t_{D(ON)I})$ . Deadtime (the denominator) has been arbitrarily held to 10% of the on-state time for a 50% duty factor. Other definitions are possible.  $t_{D(OFF)I}$  and  $t_{D(ON)I}$  are defined in Figure 17. Device turn-off delay can establish an additional frequency limiting condition for an application other than  $T_{JM}$ .  $t_{D(OFF)I}$  is important when controlling output ripple under a lightly loaded condition.

$f_{MAX2}$  is defined by  $f_{MAX2} = (P_D - P_C) / (E_{OFF} + E_{ON})$ . The allowable dissipation ( $P_D$ ) is defined by  $P_D = (T_{JM} - T_C) / R_{\theta JC}$ . The sum of device switching and conduction losses must not exceed  $P_D$ . A 50% duty factor was used (Figure 11) and the conduction losses ( $P_C$ ) are approximated by  $P_C = (V_{CE} \times I_{CE}) / 2$ .

$E_{ON}$  and  $E_{OFF}$  are defined in the switching waveforms shown in Figure 17.  $E_{ON}$  is the integral of the instantaneous power loss ( $I_{CE} \times V_{CE}$ ) during turn-on and  $E_{OFF}$  is the integral of the instantaneous power loss ( $I_{CE} \times V_{CE}$ ) during turn-off. All tail losses are included in the calculation for  $E_{OFF}$ ; i.e., the collector current equals zero ( $I_{CE} = 0$ ).

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#### NORTH AMERICA

Intersil Corporation  
P. O. Box 883, Mail Stop 53-204  
Melbourne, FL 32902  
TEL: (321) 724-7000  
FAX: (321) 724-7240

#### EUROPE

Intersil SA  
Mercure Center  
100, Rue de la Fusee  
1130 Brussels, Belgium  
TEL: (32) 2.724.2111  
FAX: (32) 2.724.22.05

#### ASIA

Intersil Ltd.  
8F-1, 96, Sec. 1, Chien-kuo North,  
Taipei, Taiwan 104  
Republic of China  
TEL: 886-2-25158508  
FAX: 886-2-25158369

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