

6 Watt Boost LED Driver



RoHS Compliance

FEATURES

- Switch Current Limit 750mA
- Drives High Voltage LED strings (38V)
- Up to 87% Efficiency
- Low Quiescent Ground Current 0.6mA
- 1MHz Fixed Frequency Low noise Operation
- Soft start "in-rush" current limiting
- Shutdown current less than 1µA
- Open LED Overvoltage Protection
- Automatic Shutdown at 1.9V (UVLO)
- Thermal overload protection
- Thin SOT23 5-Lead (1mm max height)

APPLICATIONS

- GPS Navigation Systems
- Portable Media Players
- Handheld Devices, Digital Cameras

DESCRIPTION

The CAT4240 is a DC/DC step-up converter that delivers an accurate constant current ideal for driving LEDs. Operation at a fixed switching frequency of 1MHz allows the device to be used with small value external ceramic capacitors and inductor. LEDs connected in series are driven with a regulated current set by the external resistor R1. The CAT4240 high-voltage output stage is perfect for driving mid-size and large panel displays containing up to ten white LEDs in series.

LED dimming can be done by using a DC voltage, a logic signal, or a pulse width modulation (PWM) signal. The shutdown input pin allows the device to be placed in power-down mode with "zero" quiescent current.

In addition to thermal protection and overload current limiting, the device also enters a very low power operating mode during "Open LED" fault conditions. The device is housed in a low profile (1mm max height) 5-lead thin SOT23 package for space critical applications.

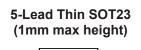
ORDERING INFORMATION

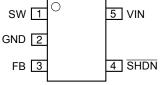
Part Number	Package	Quantity per Reel	Package Marking	
CAT4240TD-GT3	TSOT23-5 Green*	3000	TG	

* NiPdAu Plated Finish

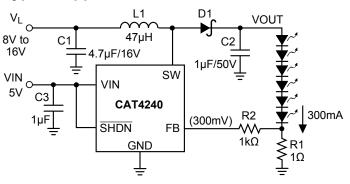
For Ordering Information details, see page 13.

Pin Configuration





Typical Application Circuit



L1: Sumida CDRH6D28-470 D1: Central CMSH1-40 (rated 40V)



ABSOLUTE MAXIMUM RATINGS

Parameters	Ratings	Units
V _{IN} , FB voltage	-0.3 to +7	V
SHDN voltage	-0.3 to +7	V
SW voltage	-0.3 to 60	V
Storage Temperature Range	-65 to +160	°C
Junction Temperature Range	-40 to +125	°C
Lead Temperature	300	°C

RECOMMENDED OPERATING CONDITIONS

Typical application circuit with external components are shown on page 1.

Parameters	Range	Units
V _{IN}	up to 5.5	V
SW pin voltage	0 to 38	V
Ambient Temperature Range ⁽¹⁾	-40 to +85	°C

DC ELECTRICAL CHARACTERISTICS

V_{IN} = 3.6V, ambient temperature of 25°C (over recommended operating conditions unless specified otherwise).

Symbol	Parameter	Test Conditions		Тур	Мах	Units	
Ι _Q	Operating Current	$V_{FB} = 0.2V$ $V_{FB} = 0.4V$ (not switching)		0.6 0.1	1.5 0.6	mA	
I _{SD}	Shutdown Current	V _{SHDN} = 0V		0.1	1	μA	
V _{FB}	FB Pin Voltage	6 LEDs with I _{LED} = 75mA	285	300	315	mV	
I _{FB}	FB pin input leakage				1	μA	
I _{LED}	Programmed LED Current	R1 = 10Ω R1 = 5Ω	28.5	30 60	31.5	mA	
V _{IH} V _{IL}	SHDN Logic High SHDN Logic Low	Enable Threshold Level Shutdown Threshold Level	0.4	0.8 0.7	1.5	V V	
F _{SW}	Switching Frequency		0.8	1.0	1.3	MHz	
DC	Maximum Duty Cycle	V _{IN} = 3V		92		%	
I _{LIM}	Switch Current Limit	V _{IN} = 3.6V V _{IN} = 5V	600 750			mA	
R _{sw}	Switch "On" Resistance	I _{sw} = 100mA		1.0	2.0	Ω	
I _{LEAK}	Switch Leakage Current	Switch Off, V _{SW} = 30V		2	5	μA	
	Thermal Shutdown			150		°C	
	Thermal Hysteresis			20		°C	
V _{UVLO}	Undervoltage Lockout (UVLO) Threshold			1.9		V	
V _{OV-SW}	Overvoltage Detection Threshold			40		V	
V _{OCL}	Output Voltage Clamp	"Open LED"		42		V	

Note:

(1) Thin SOT23-5 package thermal resistance θ_{JA} = 135°C/W when mounted on board over a ground plane.



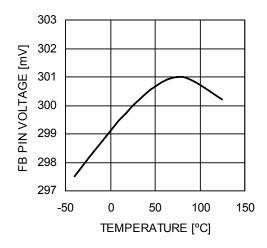
TYPICAL CHARACTERISTICS

VIN = 5V, V_L = 13V, T_{AMB} = 25°C, typical application circuit unless otherwise specified.

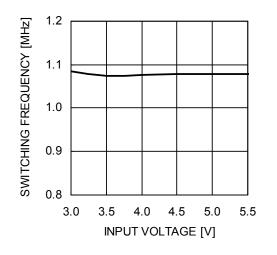
$\begin{array}{c} 200 \\ V_{FB} = 0.4V \\ 150 \\ 100 \\ 50 \\ 3.0 \\ 3.5 \\ 4.0 \\ 4.5 \\ 5.0 \\ 5.5 \\ INPUT VOLTAGE [V] \end{array}$

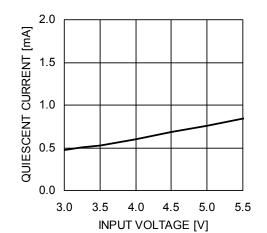
Quiescent Current vs. VIN (Not Switching)

FB pin voltage vs. Temperature



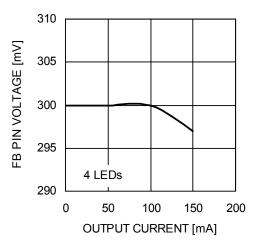
Switching Frequency vs. Supply Voltage



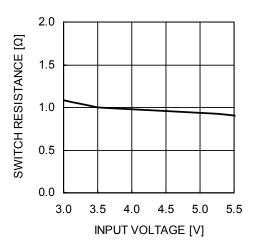


Quiescent Current vs. VIN (Switching)

FB Pin Voltage vs. Output Current



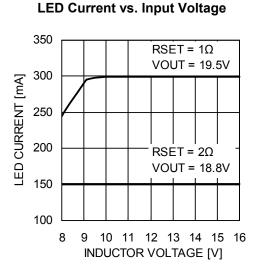
Switch ON Resistance vs. Input Voltage



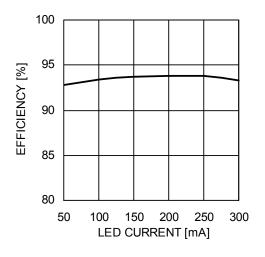


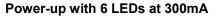
TYPICAL CHARACTERISTICS

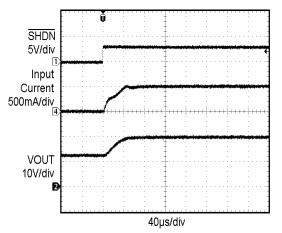
VIN = 5V, V_L = 13V, T_{AMB} = 25°C, typical application circuit unless otherwise specified.



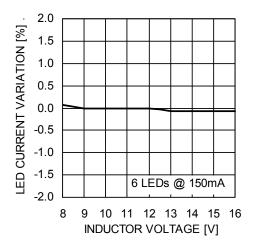
Efficiency vs. Load Current (6 LEDs)



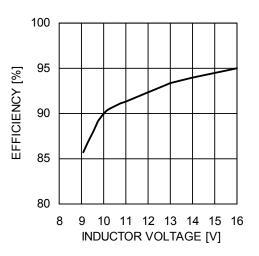




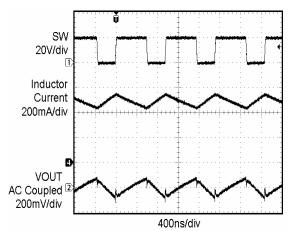
LED Current Regulation



Efficiency vs. Inductor Voltage (6 LEDs)



Switching Waveform





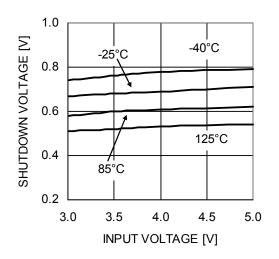
TYPICAL CHARACTERISTICS

VIN = 5V, V_L = 13V, T_{AMB} = 25°C, typical application circuit unless otherwise specified.

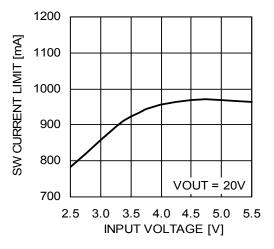
500 400 VOUT = 20V 300 200 100 8 9 10 11 12 13 14 15 16 INDUCTOR VOLTAGE [V]

Maximum Output Current

Shutdown Voltage



Switch Current Limit





PIN DESCRIPTION

VIN is the supply input for the internal logic. The device is compatible with supply voltages down to 2.8V and up to 5.5V. It is recommended that a small bypass ceramic capacitor $(4.7\mu F)$ be placed between the VIN and GND pins near the device. If the supply voltage drops below 1.9V, the device stops switching.

SHDN is the shutdown logic input. When the pin is tied to a voltage lower than 0.4V, the device is in shutdown mode, drawing nearly zero current. When the pin is connected to a voltage higher than 1.5V, the device is enabled.

GND is the ground reference pin. This pin should be connected directly to the ground plane on the PCB.

SW pin is connected to the drain of the internal CMOS power switch of the boost converter. The inductor and the Schottky diode anode should be connected to the SW pin. Traces going to the SW pin should be as short as possible with minimum loop area. An overvoltage detection circuit is connected to the SW pin. When the voltage reaches 40V, the device enters a low power operating mode preventing the SW voltage from exceeding the maximum rating.

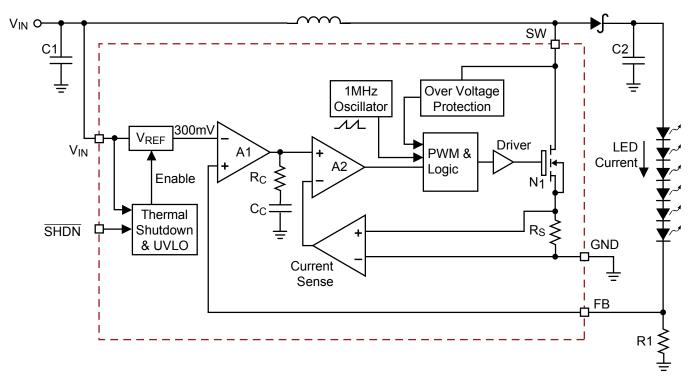
FB feedback pin is regulated at 0.3V. A resistor connected between the FB pin and ground sets the LED current according to the formula:

$$I_{LED} = 0.3 V/R1$$

The lower LED cathode is connected to the FB pin.

Pin #	Name	Function
1	SW	Switch pin. This is the drain of the internal power switch.
2	GND	Ground pin. Connect the pin to the ground plane.
3	FB	Feedback pin. Connect to the last LED cathode.
4	SHDN	Shutdown pin (Logic Low). Set high to enable the driver.
5	VIN	Power Supply input.

BLOCK DIAGRAM





DEVICE OPERATION

The CAT4240 is a fixed frequency (1MHz), low noise, inductive boost converter that provides a constant current with excellent line and load regulation. The device uses a high-voltage CMOS power switch between the SW pin and ground to energize the inductor. When the switch is turned off, the stored energy in the inductor is released into the load via the Schottky diode.

The on/off duty cycle of the power switch is internally adjusted and controlled to maintain a constant regulated voltage of 0.3V across the feedback resistor connected to the feedback pin (FB). The value of the resistor sets the LED current accordingly (0.3V/R1).

During the initial power-up stage, the duty cycle of the internal power switch is limited to prevent excessive in-rush currents and thereby provide a "soft-start" mode of operation.

When the inductor is connected to a 9V supply or higher, the CAT4240 can drive 6 LEDs in series at

300mA delivering a total power of 6 Watts into the load. A separate 5V supply voltage is connected to the VIN pin.

In the event of an "Open LED" fault condition, where the feedback control loop becomes open, the output voltage will continue to increase. Once this voltage exceeds 40V, an internal protection circuit will become active and place the device into a very low power safe operating mode where only a small amount of power is transferred to the output. This is achieved by pulsing the switch once every $60\mu s$ or so, and keeping it on for about $1\mu s$.

Thermal overload protection circuitry has been included to prevent the device from operating at unsafe junction temperatures above 150°C. In the event of a thermal overload condition the device will automatically shutdown and wait till the junction temperatures cools to 130°C before normal operation is resumed.



APPLICATION INFORMATION

External Component Selection

Capacitors

The CAT4240 only requires small ceramic capacitors of 4.7μ F on the inductor input, 1μ F on the VIN pin and 1μ F on the output. Under normal condition, a 4.7μ F input capacitor is sufficient. For applications with higher output power, a larger input capacitor of 10μ F may be appropriate. X5R and X7R capacitor types are ideal due to their stability across temperature range.

Inductor

A 47μ H inductor is recommended for most of the CAT4240 applications. In cases where the efficiency is critical, inductances with lower series resistance are preferred. Inductors with current rating of 800mA or higher are recommended for most applications. Sumida CDRH6D28-470 47μ H inductor has a rated current of 800mA and a series resistance (D.C.R.) of 176m Ω typical.

Schottky Diode

The current rating of the Schottky diode must exceed the peak current flowing through it. The Schottky diode performance is rated in terms of its forward voltage at a given current. In order to achieve the best efficiency, this forward voltage should be as low as possible. The response time is also critical since the driver is operating at 1MHz. Central Semiconductor Schottky rectifier CMSH1-40 (1A rated) is recommended for most applications.

LED Current Setting

The LED current is set by the external resistor R1 connected between the feedback pin (FB) and ground. The formula below gives the relationship between the resistor and the current:

R1 = 0.3V	LED current
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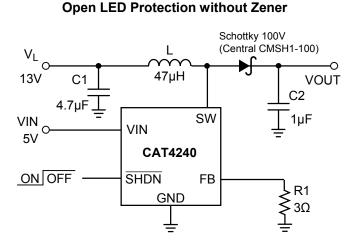
LED current (mA)	R1 (Ω)
20	15
25	12
30	10
100	3
300	1

Table 1. Resistor R1 and LED current

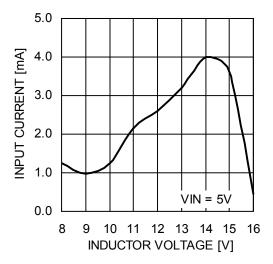


OPEN LED PROTECTION

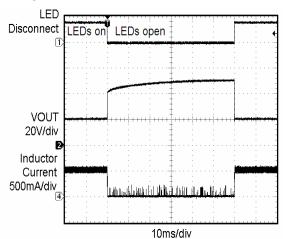
In the event of an "Open LED" fault condition, the CAT4240 will continue to boost the output voltage with maximum power until the output voltage reaches approximately 40V. Once the output exceeds this level, the internal circuitry immediately places the device into a very low power mode where the total input power is limited to about 6mW (about 1.6mA input current with a 3.6V supply). The SW pin clamps at a voltage below its maximum rating of 60V. There is no need to use an external zener diode between Vout and the FB pin. A 50V rated C2 capacitor is required to prevent any overvoltage damage in the open LED condition.

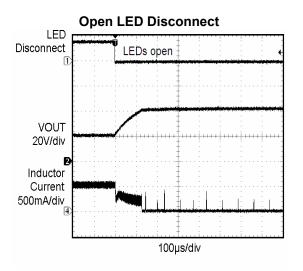


Open LED Supply Current vs. VIN without Zener

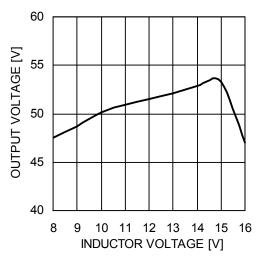


Open LED Disconnect and Reconnect





Open LED Output Voltage vs. VIN without Zener



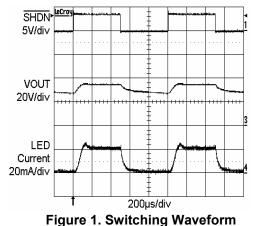


DIMMING CONTROL

There are several methods available to control the LED brightness.

PWM signal on the SHDN Pin

LED brightness dimming can be done by applying a PWM signal to the SHDN input. The LED current is repetitively turned on and off, so that the average current is proportional to the duty cycle. A 100% duty cycle, with SHDN always high, corresponds to the LEDs at nominal current. Figure 1 shows a 1kHz signal with a 50% duty cycle applied to the SHDN pin. The recommended PWM frequency range is from 100Hz to 2kHz.



with 1kHz PWM on SHDN

Filtered PWM Signal

A filtered PWM signal used as a variable DC voltage can control the LED current. Figure 2 shows the PWM control circuitry connected to the CAT4240 FB pin. The PWM signal has a voltage swing of 0V to 2.5V. The LED current can be dimmed within a range from 0mA to 20mA. The PWM signal frequency can vary from very low frequency up to 100kHz.

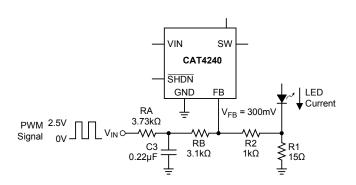
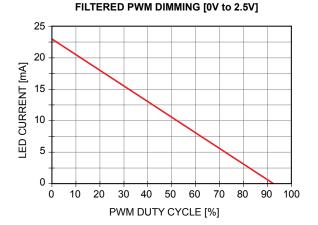


Figure 2. Circuit for Filtered PWM Signal

A PWM signal at 0V DC, or a 0% duty cycle, results in a max LED current of about 22mA. A PWM signal with a 93% duty cycle or more, results in an LED current of 0mA.





BOARD LAYOUT

The CAT4240 is a high-frequency switching regulator. The traces that carry the high-frequency switching current have to be carefully laid out on the board in order to minimize EMI, ripple and noise in general. The thicker lines on Figure 3 show the switching current path. All these traces have to be short and wide enough to minimize the parasitic inductance and resistance. The loop shown on Figure 3 corresponds to the current path when the CAT4240 internal switch is closed. On Figure 4 is shown the current loop, when the CAT4240 switch is open. Both loop areas should be as small as possible. Capacitor C1 has to be placed as close as possible to the VIN pin and GND. The capacitor C2 has to be connected separately to the top LED anode. A ground plane under the CAT4240 allows for direct connection of the capacitors to ground. The resistor R1 must be connected directly to the GND pin of the CAT4240 and not shared with the switching current loops and any other components.

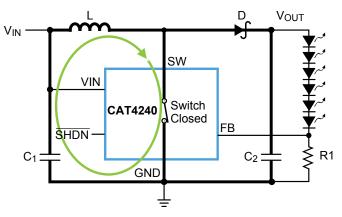


Figure 3. Closed-switch Current Loop

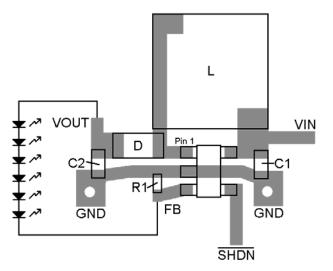


Figure 5. Recommended PCB Layout

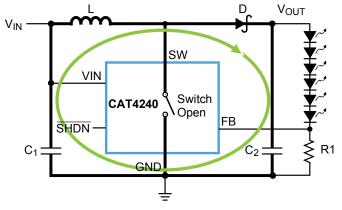
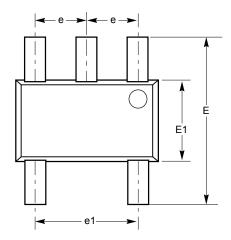
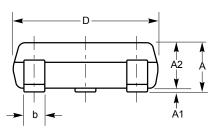


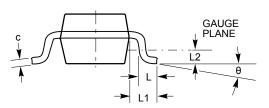
Figure 4. Open-switch Current Loop



PACKAGE OUTLINES 5-LEAD THIN SOT-23







SYMBOL	MIN NOM		MAX	
А	—	—	1.0	
A1	0.01	0.05	0.1	
A2	0.80	0.87	0.9	
b	0.30	—	0.45	
с	0.12	0.15	0.20	
D	2.90 BSC			
E	2.80 BSC			
E1	1.60 BSC			
е	1.95 BSC			
e1	1.90 BSC			
L	0.30 0.40 0.50			
L1	0.60 REF			
L2	0.25 BSC			
q	0° 8°			

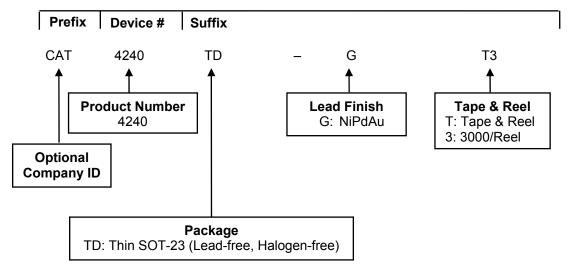
Notes:

(1) All dimensions are in millimeters, angles in degrees.

(2) Refer JEDEC MO-193.



EXAMPLE OF ORDERING INFORMATION



Notes:

- (1) All packages are RoHS-compliant (Lead-free, Halogen-free).
- (2) The standard lead finish is NiPdAu.
- (3) The device used in the above example is a CAT4240–GT3 (TSOT-23, NiPdAu, Tape & Reel).
- (4) For additional package and temperature options, please contact your nearest Catalyst Semiconductor Sales office.

SERIES LED DRIVERS

Part Number	V _{IN} [V]	LEDs	V _{FB} [mV]	I _{SW-LIM} [mA] ⁽¹⁾	V _{оυт} (max) [V]	Dimming Interface	Packages
CAT4137	2.0 - 5.5	5	300	350	24	PWM	TSOT23–5
CAT4237	2.0 - 5.5	8	300	450	30	PWM	TSOT23–5
CAT4238	2.0 - 5.5	10	300	450	38	PWM	TSOT23–5
CAT4240	2.0 - 5.5	10	300	750	38	PWM	TSOT23–5

Note:

(1) Switch current limit, typical values

REVISION HISTORY

Date	Rev.	Reason
05/31/2007	Α	Initial Issue

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